



Contract Number  
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# Ultra-High Temperature Distributed Sensors

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# Agenda



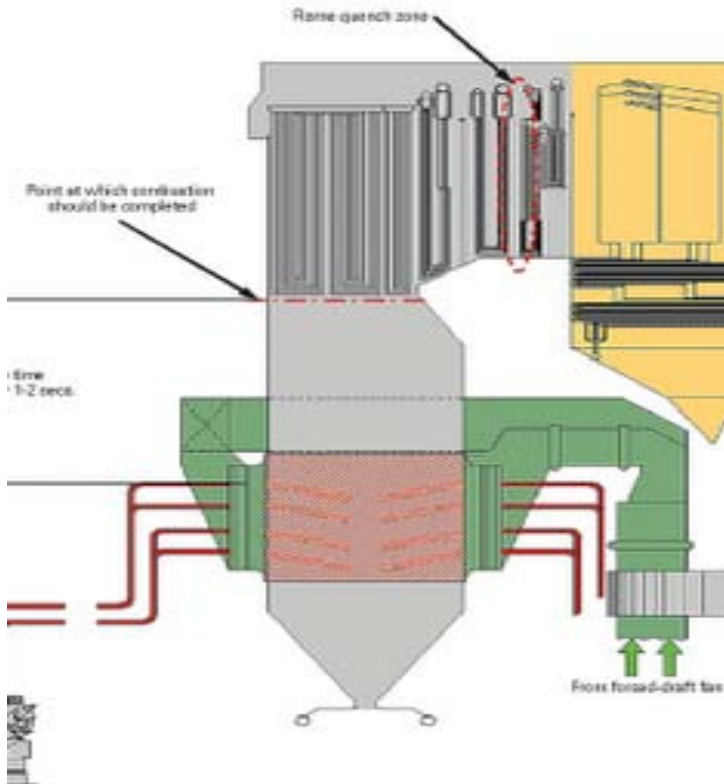
- Problem
- Metamaterial Solution
  - Design
  - Materials characterization
  - Prototype Testing
- Mechanically Modulated Antenna Solution
  - System Design
  - Prototype Testing
  - Antenna Design
  - Temperature Tests
- Characterization of RF Environment
- Next Steps

# The Problem

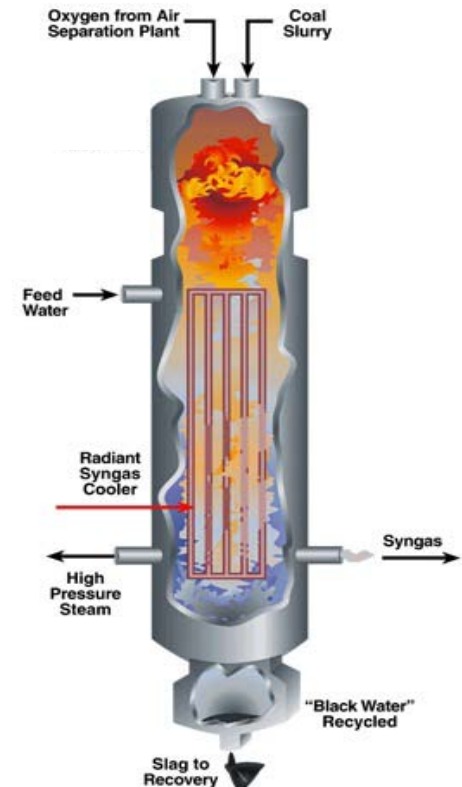


To measure physical parameters (temperature, pressure, strain, etc.) in extreme temperature and highly corrosive environments.

## Coal-Fired Boiler Plant

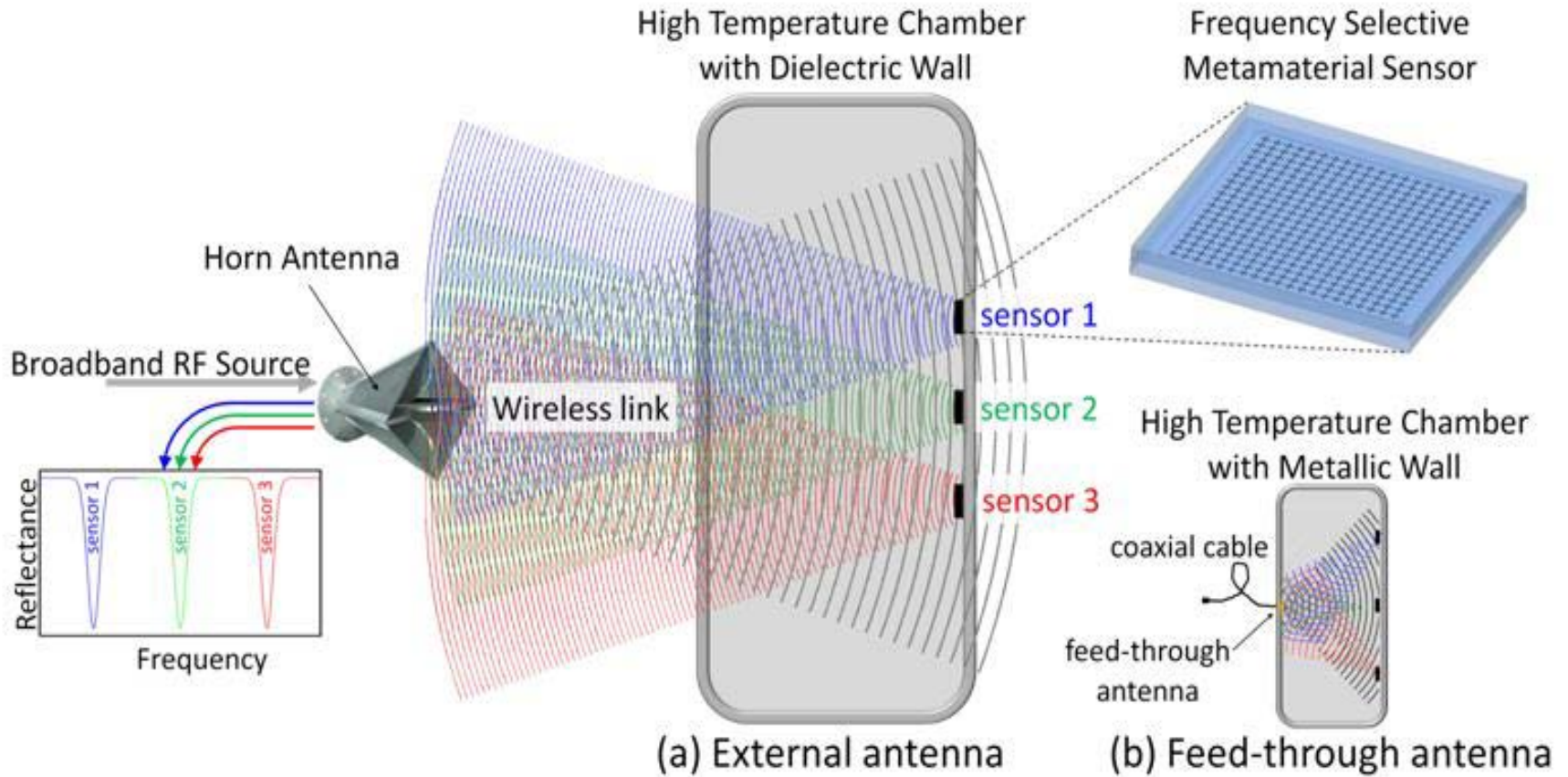


## Coal-Gasification Plant



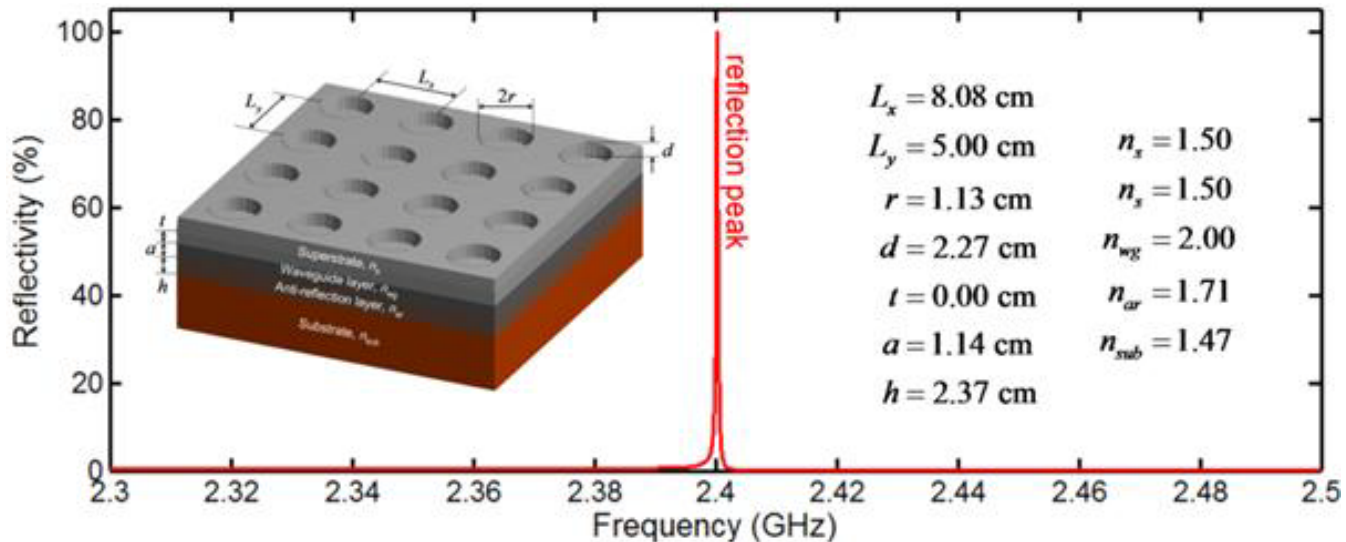
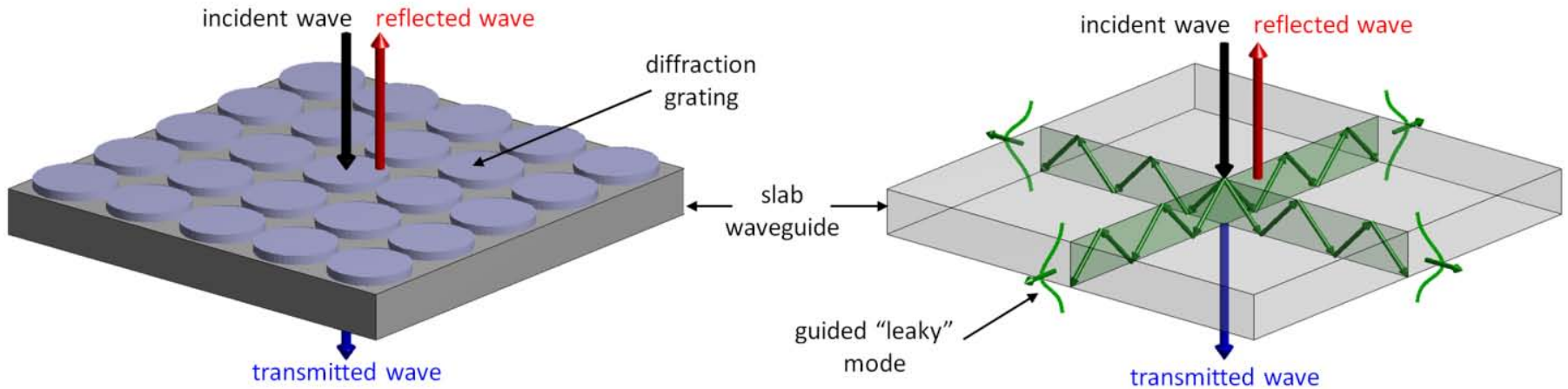
# The Proposed Solution

## Passive Wireless Sensors



# The Initial Approach

## Guided Mode Resonance Filters (GMRFs)

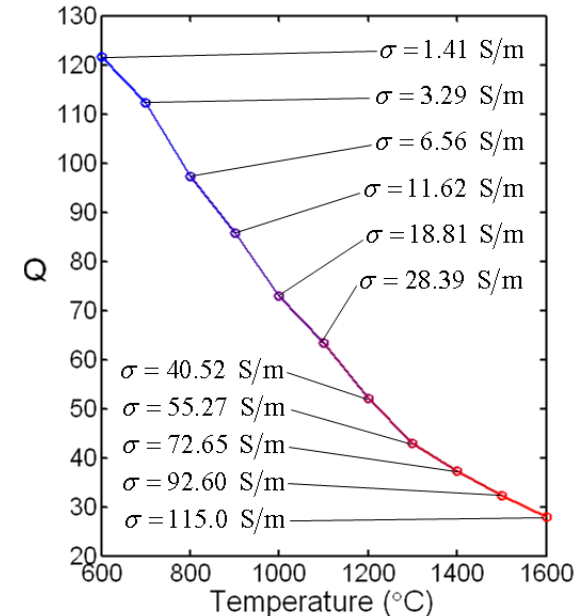
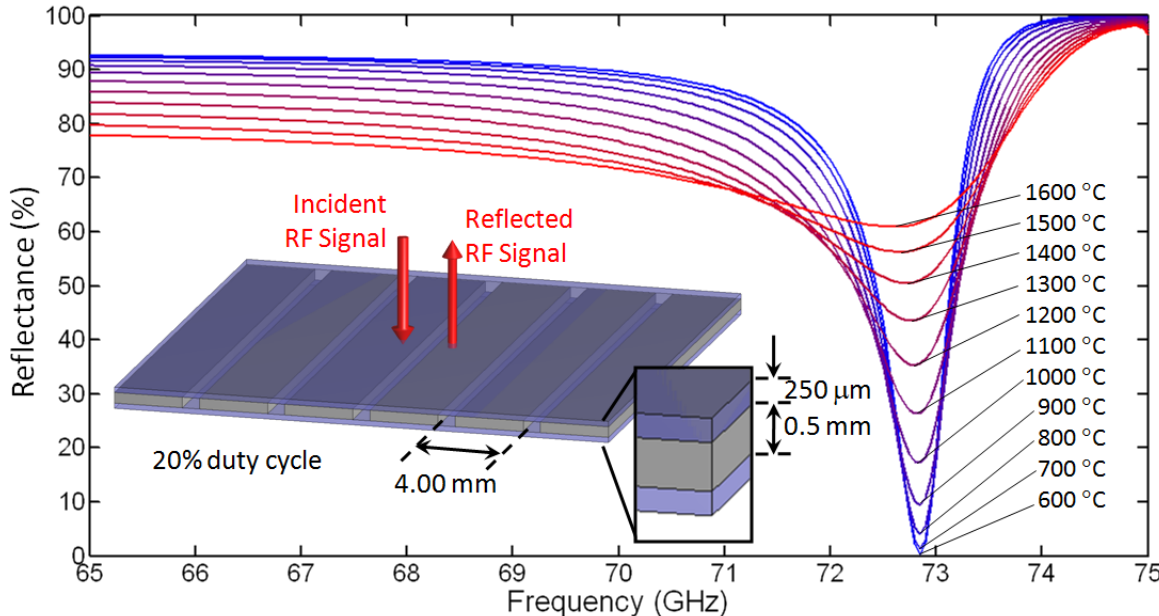
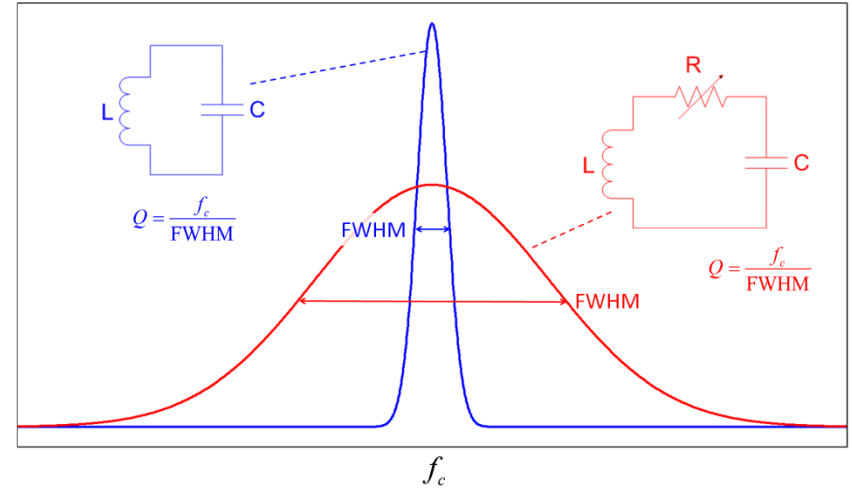


# Initial Approach

## Transduction concept



- Periodic arrays on the order of the RF wavelength generate a signature
- RF material properties change with temperature
- Signature changes can be calibrated to temperature
- Sensors can be designed to have distinct signatures enabling multiplex



# Materials Characterization Overview



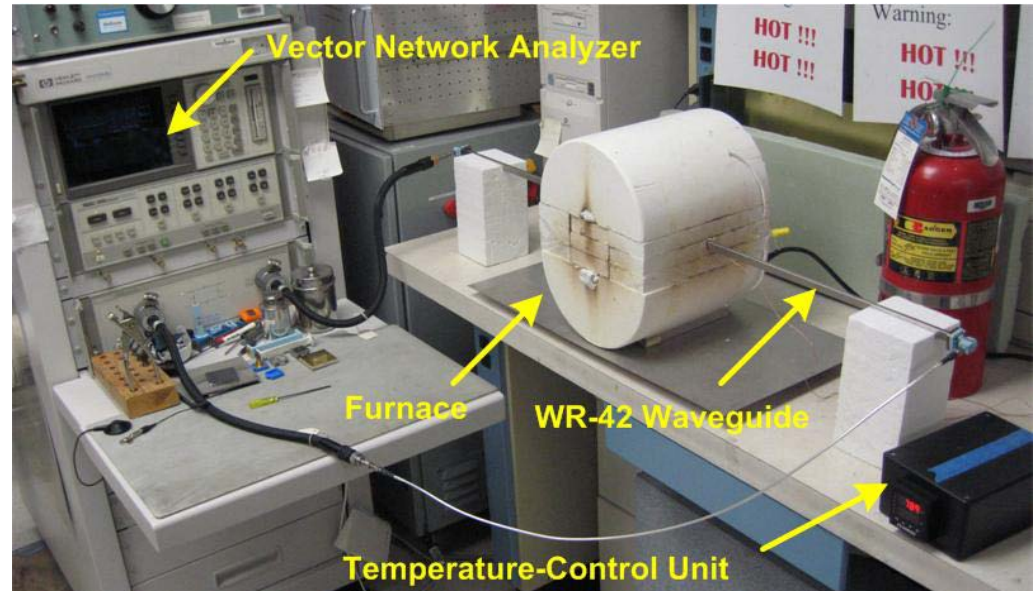
Goal: Quantify temperature dependent RF material properties

- permittivity ( $\epsilon$ ) and permeability ( $\mu$ )
- measure transmission and reflection thru samples to back out the conductivity

- $\epsilon_r$  is the relative permittivity
- $\mu_r$  is the relative permeability
- $\tan \delta_\epsilon$  and  $\tan \delta_\mu$  are electric and magnetic loss tangents respectively
- $\sigma_e$  is dielectric conductivity, and  $\omega$  is angular frequency

$$\mu = \mu_0(\mu_r' - j\mu_r'') = \mu_0\mu_r'(1 - j \tan \delta_\mu)$$

$$\epsilon = \epsilon_0\epsilon_r' - j\frac{\sigma_e}{\omega}$$



# Materials Characterization

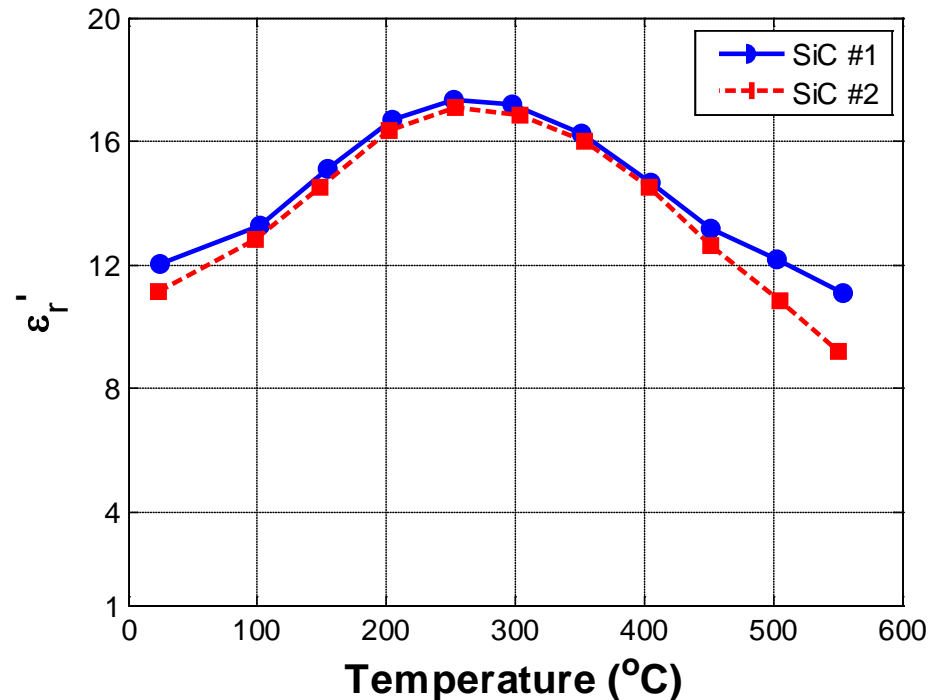
## Experimental results - SiC



- Dielectric constant and Loss computed from raw data ( $S_{11}$ ,  $S_{21}$ )
- loss increases significantly with temperature

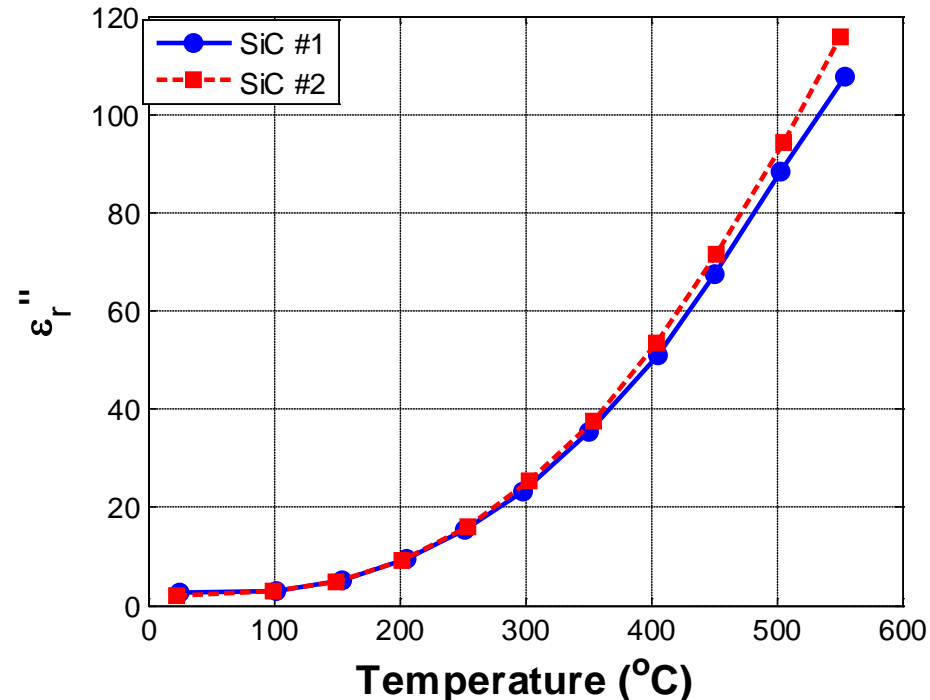
### Dielectric Constant ( $\epsilon_r'$ )

22 GHz



### Loss ( $\epsilon_r''$ )

22 GHz



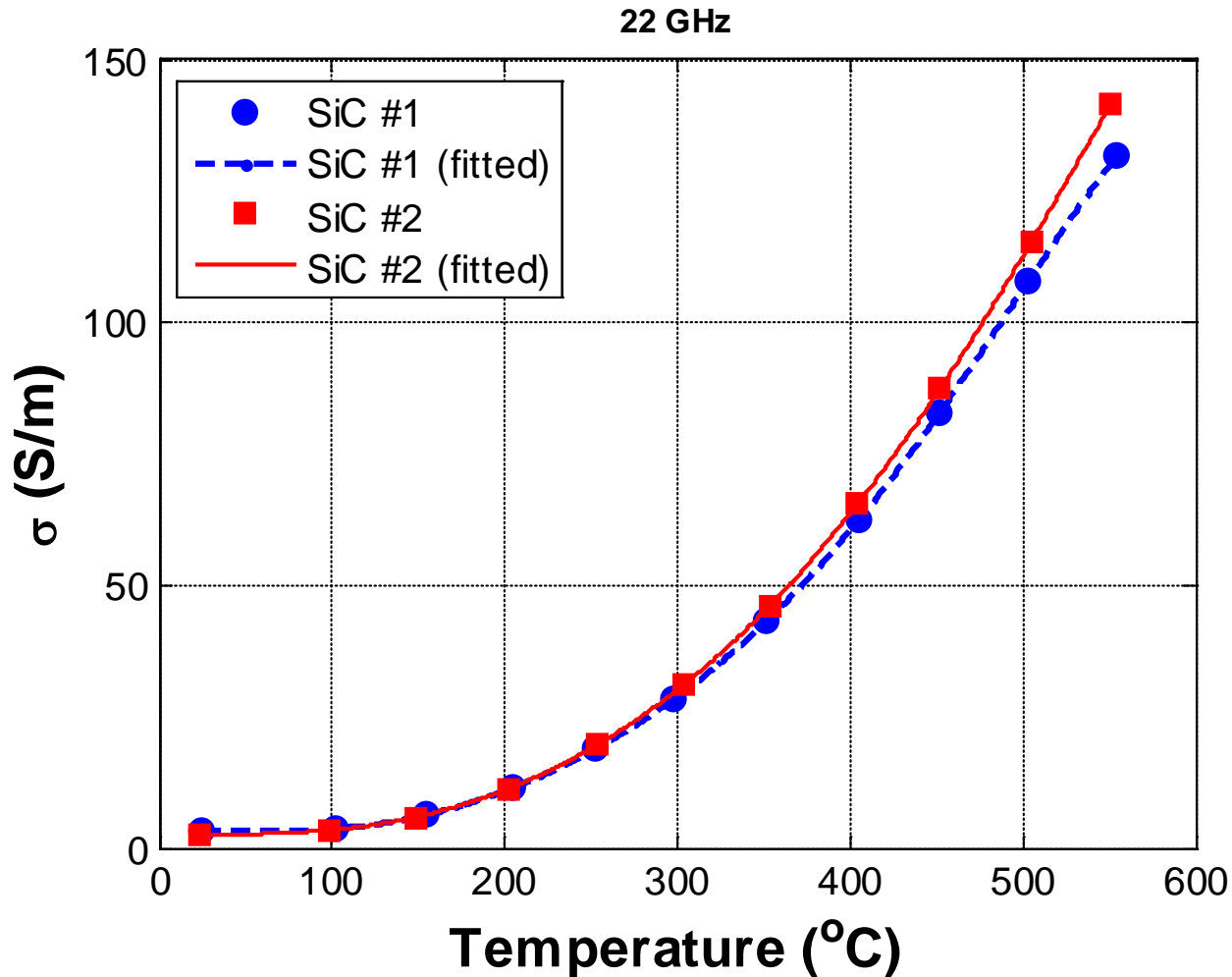


# Materials Characterization

## *Experimental results*



Conductivity vs temperature for SiC at 22 GHz



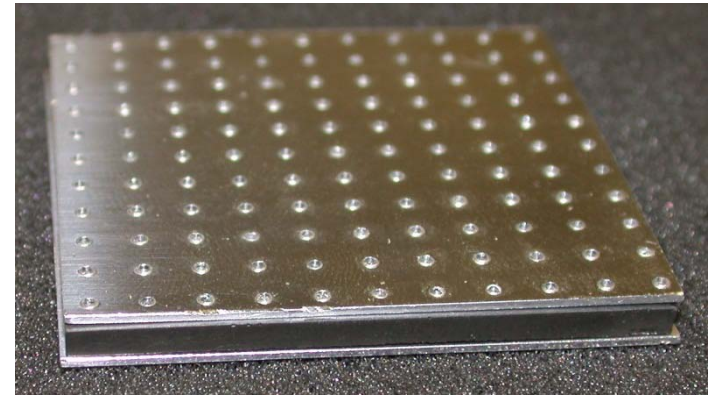
# Sensor Design

## *GMR sensor concept*

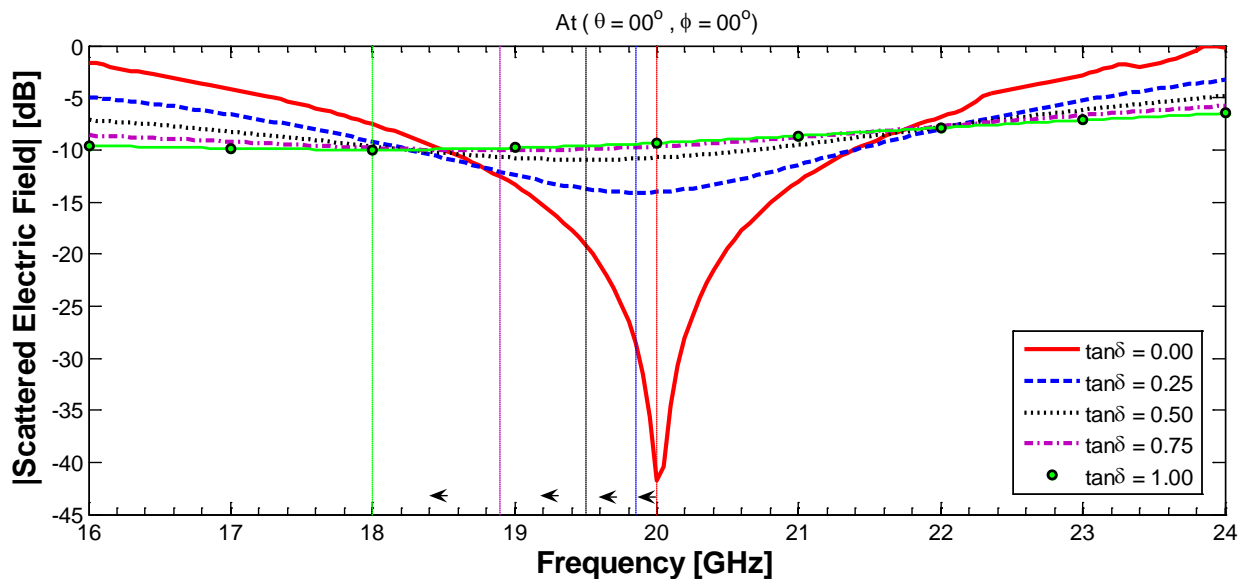


### Guided Mode Resonance (GMR)

- incident energy at a specific frequency is couple into a transverse longitudinal mode, creating a notch in the reflected response
- simple construction
- sensitive to angle of incidence



Guided mode sensor



# Sensor Design

## *Alternative sensor concepts*



### Element Resonance (SRR, dipole array etc.)

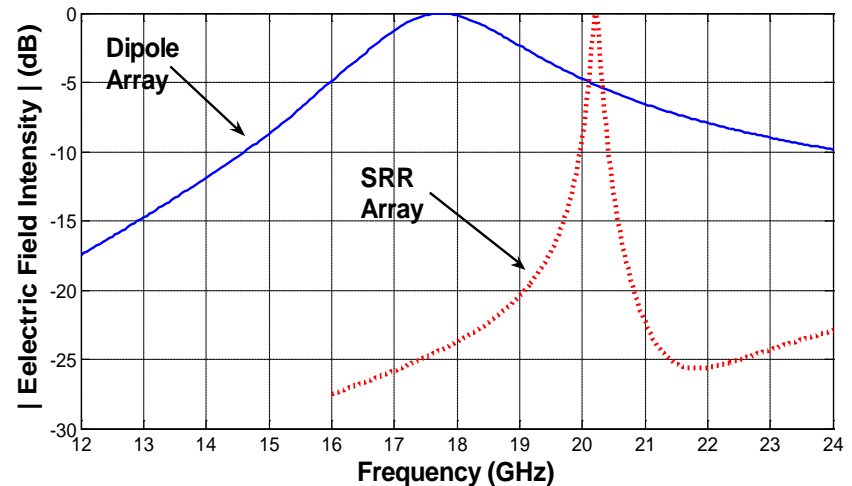
- less angle sensitive
- more complex geometry



Dipole array



Split ring resonator



# Sensor Design

## *Sensor testing*



- room temperature evaluation
- baseline measurements
- validation of test setup

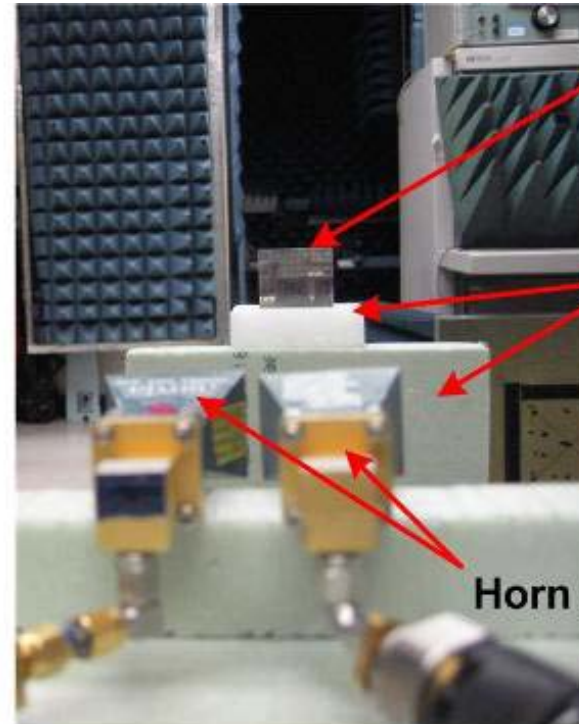


HP 8510C VNA



Styrofoam

Horn Antennas



Target

Styrofoam

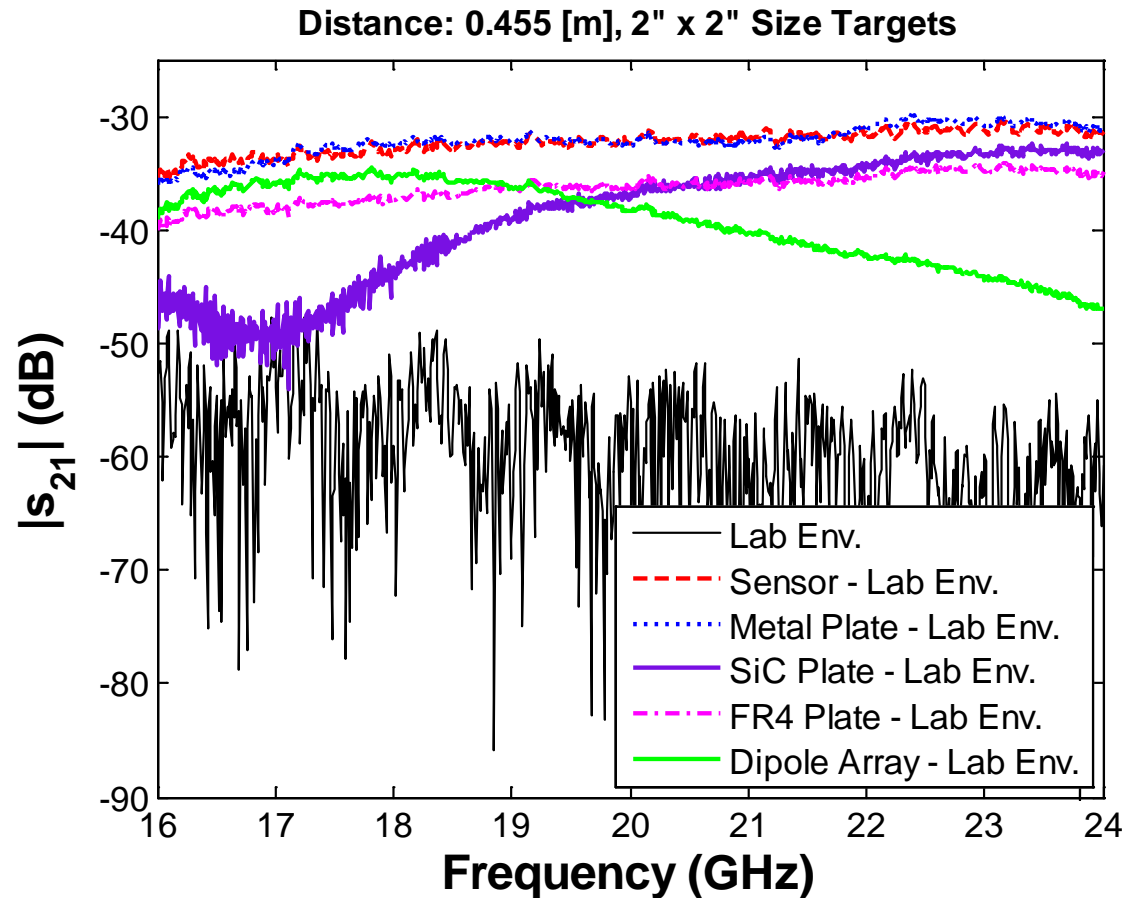
Horn Antennas

# Sensor Design

## *Sensor testing*



- measurement of reflected energy vs frequency
- 16 – 24 GHz
- baseline data for “no target”
- dipole array shows good comparison to simulation
- GMR sensor has not shown simulated response
- edge diffraction effects possibly dominating response
- high loss of SiC results in weak coupling

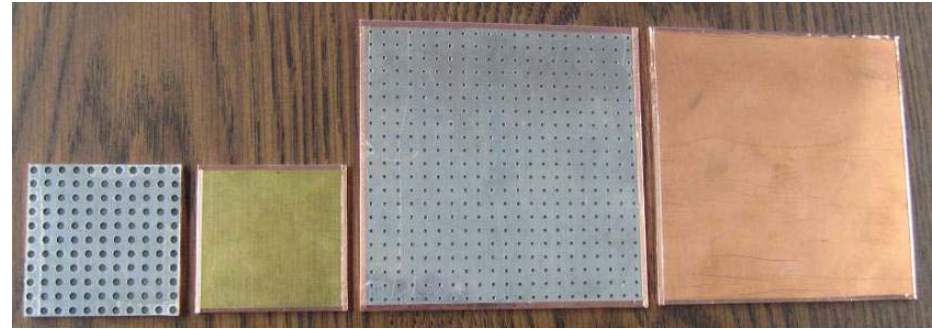


# Sensor Design

## *Sensor testing*

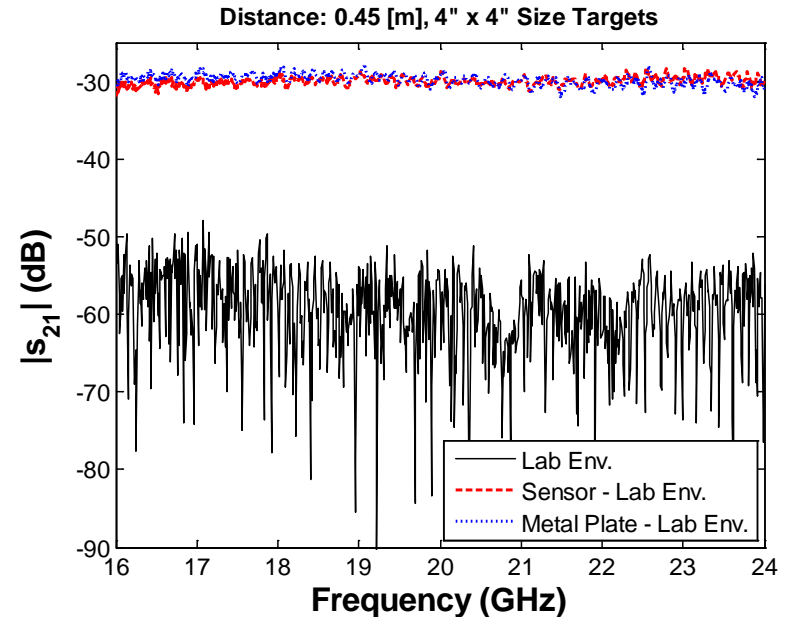
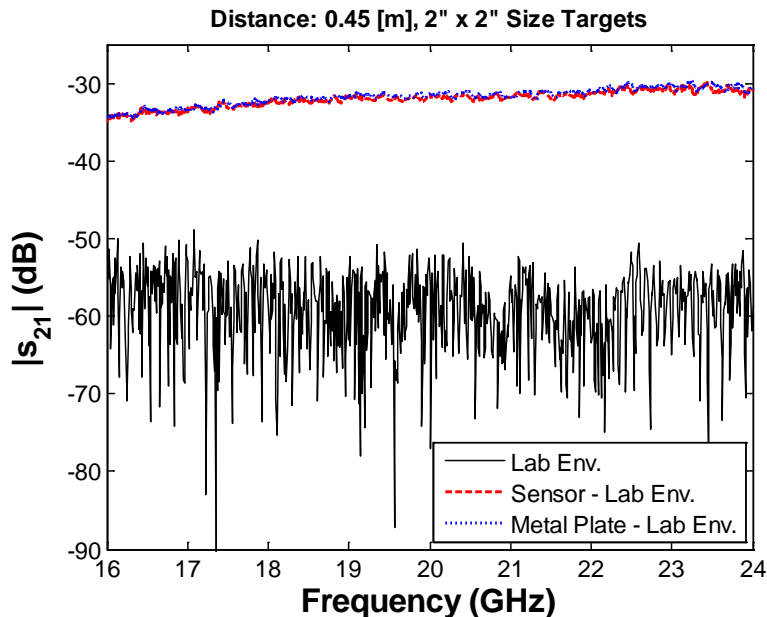


- Measurement of reflected energy vs frequency
- 16 – 24 GHz
- baseline data for “no target”
- Metamaterial sensor response compared to a metal plate
- Sensor response is not discernable, even at normal incidence



2" x 2"

4" x 4"



# Sensor Design

## *Challenges with GMR (Metamaterial) Approach*



1) *The sensor response was weak and indistinguishable from background clutter.*

- Surface reflections swamped the GMR response
- Material dielectric loss may prevent good coupling of the guided mode

⇒ Need a unique signature to discriminate sensor signal from background reflections (frequency shifting, harmonics, etc.)

2) *A working metamaterial sensor would have to be very large to have a sufficient number of periods*

- Even a 4" x 4" sensor (approx 20 periods) had a response dominated by surface reflections and edge effects

⇒ Need a device that does not rely upon periodic structures of integral wavelength (single element devices)

3) *The sensor was angle sensitive.*

- The sensor response changed or disappeared with any deviation from exact normal (90 degree) angle of incidence

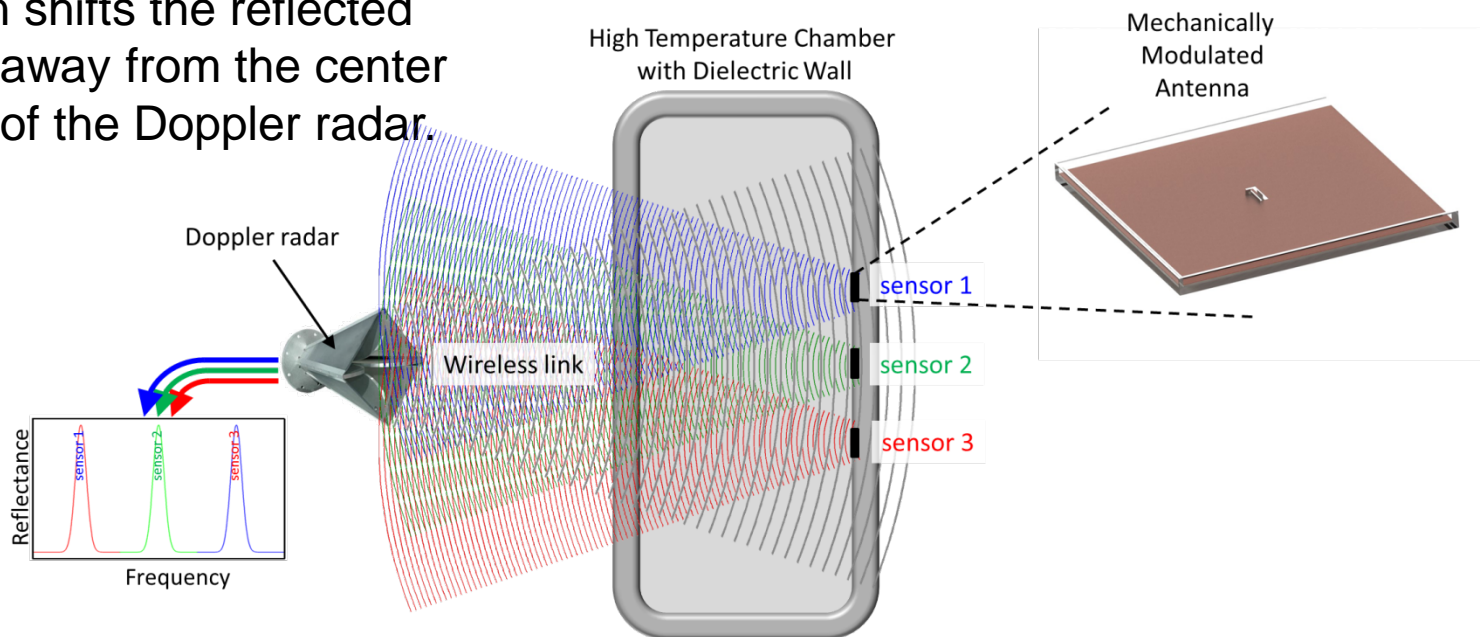
⇒ Need a device that does not rely upon normal incidence of interrogating radio wave.

# A New Wireless Sensor Design

## *Mechanically Modulated Antenna*

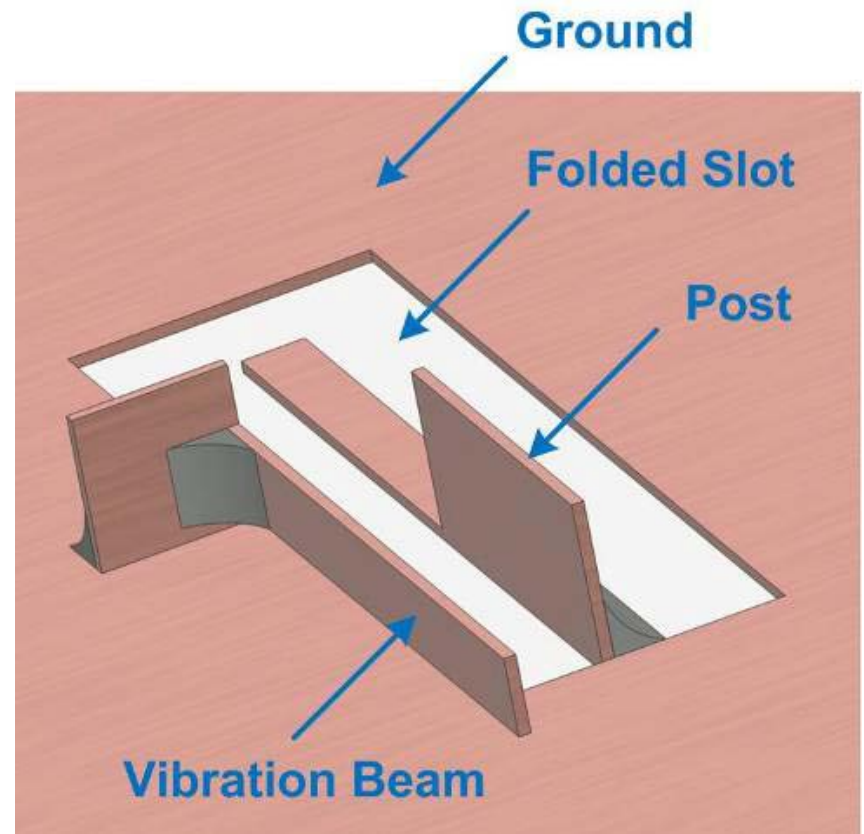
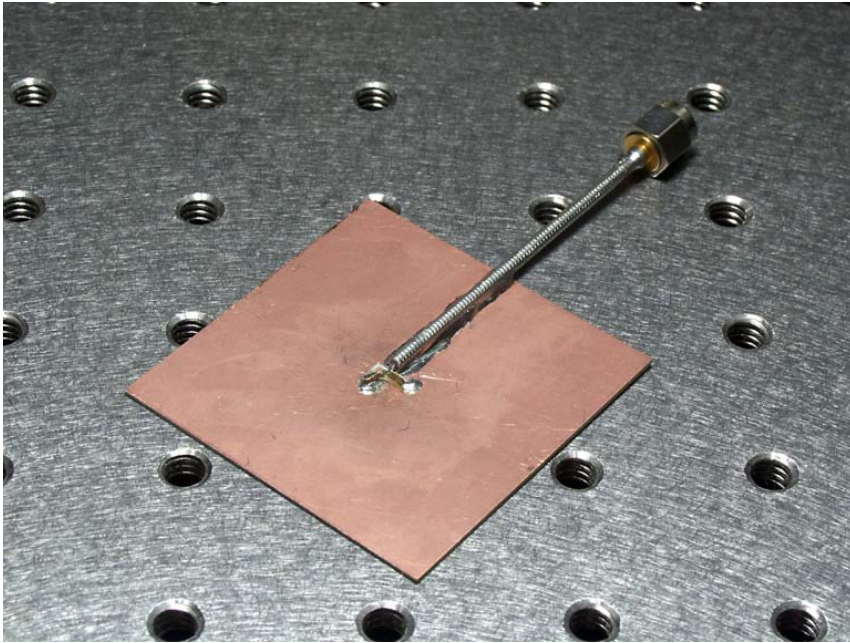


- Each sensor is an antenna with a mechanical element which is interrogated wirelessly by Doppler radar.
- The mechanical element modulates the frequency response of the antenna.
  - The modulation frequency changes with temperature of the antenna.
  - Modulation shifts the reflected frequency away from the center frequency of the Doppler radar.
- Each antenna can be tuned to a unique frequency to multiplex several sensors with a single interrogator (radar)
- Each sensor has a unique signature outside spectral range of background reflections

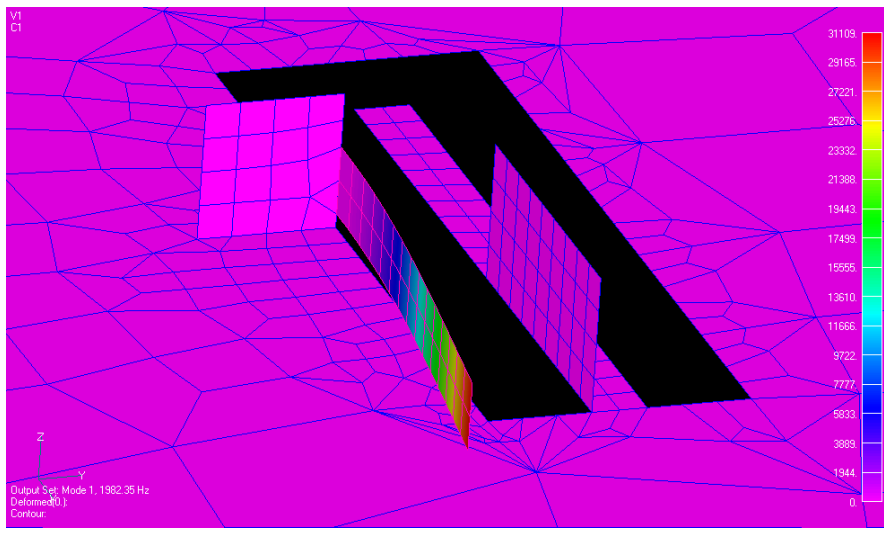
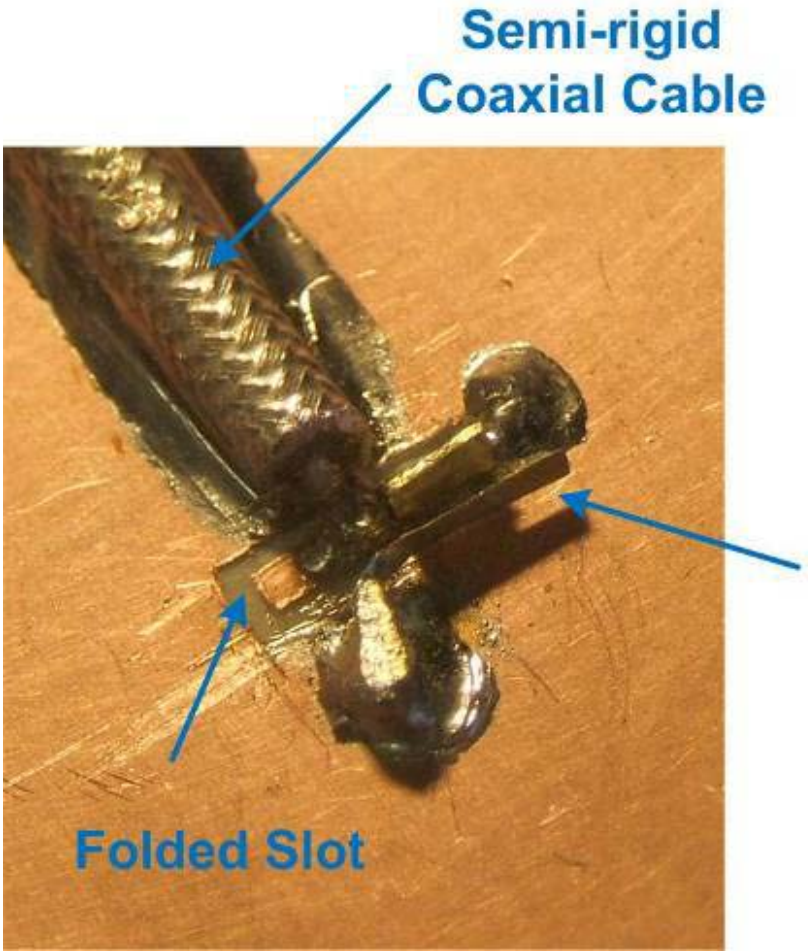




# Initial Sensor Geometry



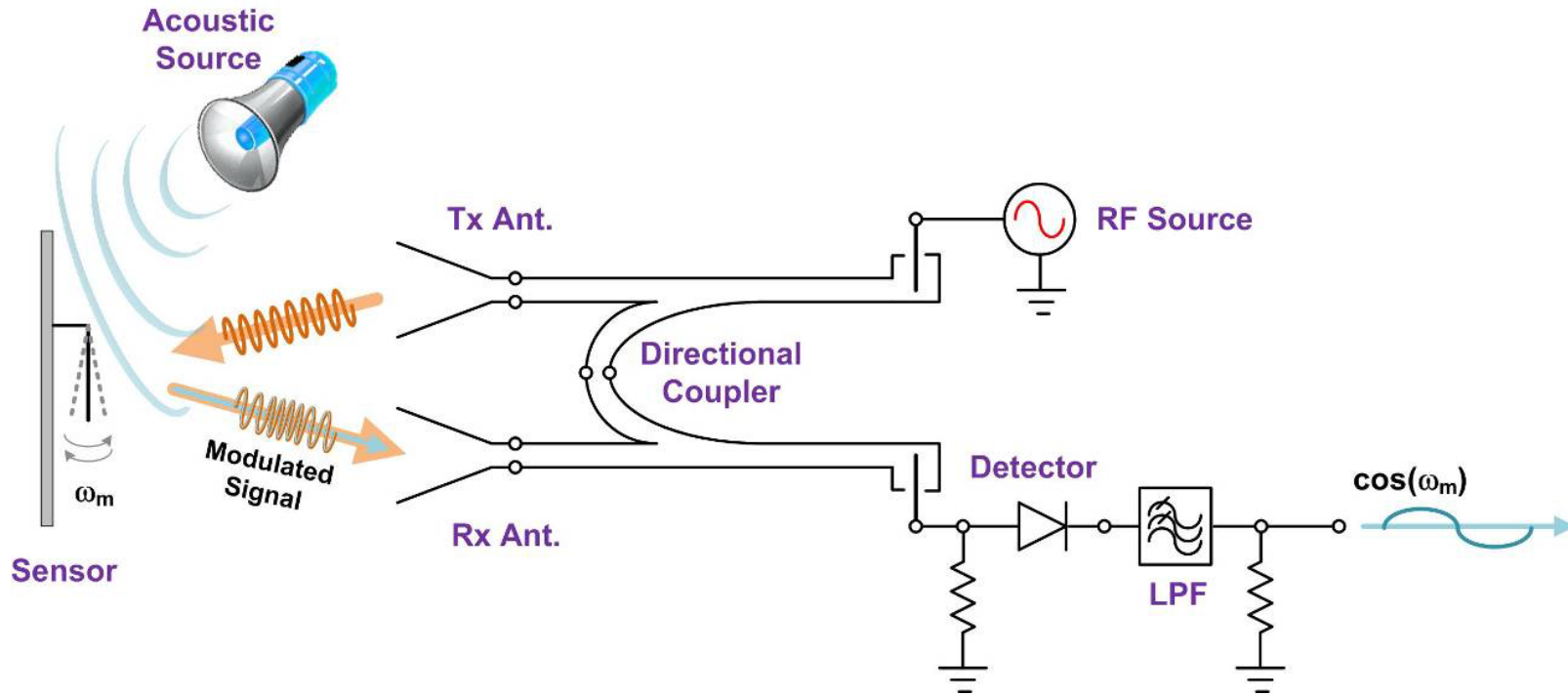
# Slot Antenna Close-Up



FEA predicts beam natural frequency of 1980 Hz.

# Interrogator Design

## Doppler RADAR

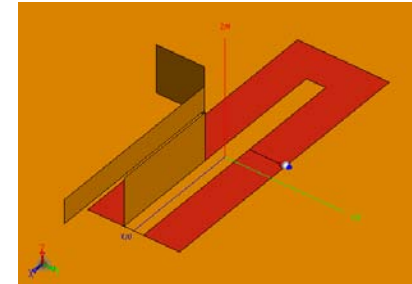
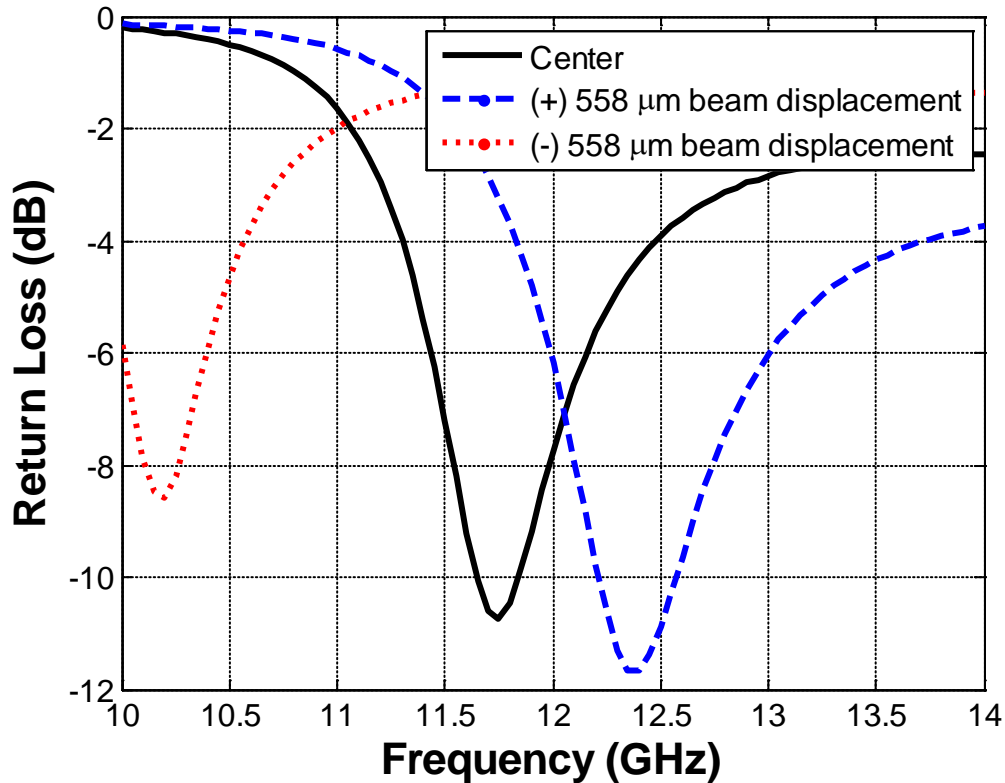


- Acoustic excitation (1800 Hz)
- Doppler RF interrogation (10 GHz)

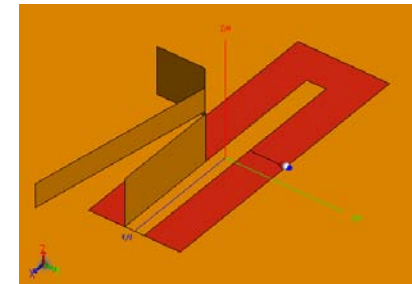
# FEKO Simulation Results



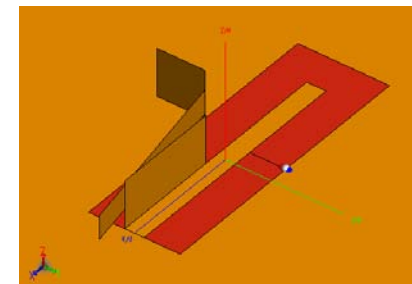
Beam displacement changes antenna impedance and modulates antenna center frequency.



center



+ 558 μm



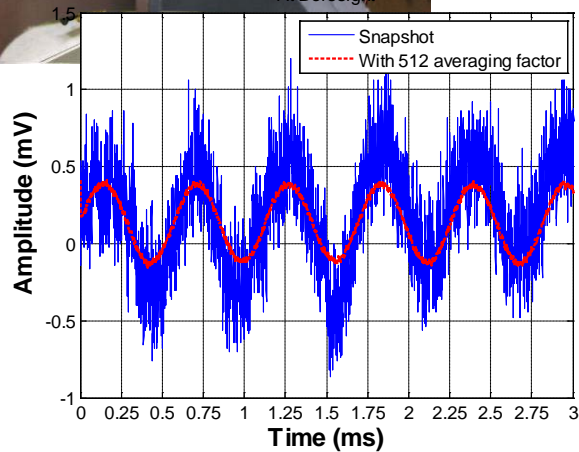
- 558 μm

# Experimental Results

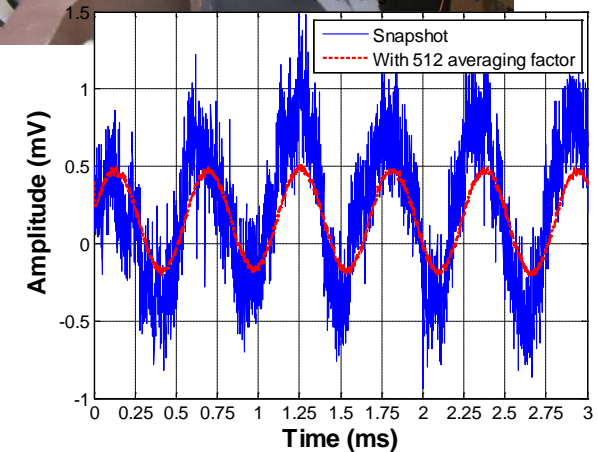
## *Mechanically Modulated Antenna*



Normal incidence



70° incidence



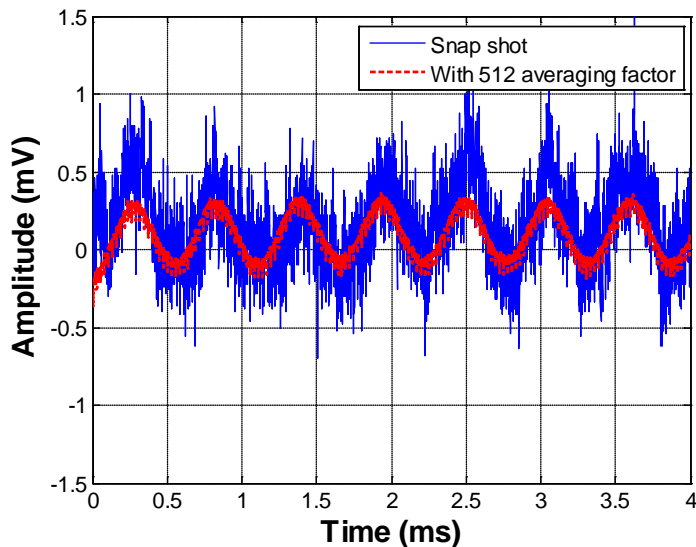
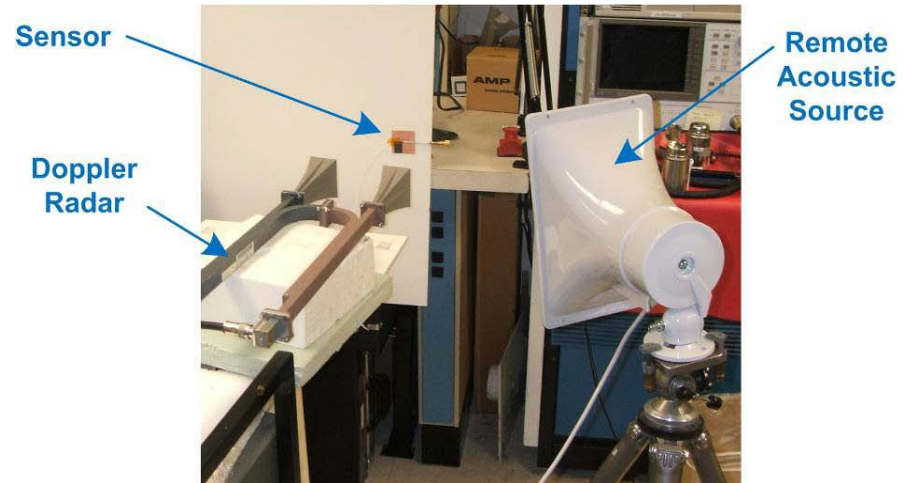
Initial experiments used laser excitation

# MMA Experimental Results

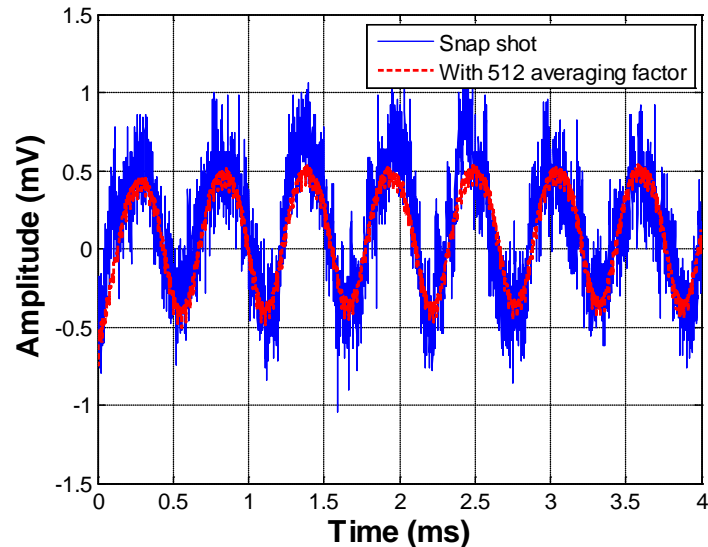
## *Acoustic Excitation*



- RF Doppler placed 30 cm from target with 15 dBm power (30 mW)
- Acoustic source 1 m from target with 110 dBA at 1800 Hz
- Better acoustic capture efficiency when mounted on cardboard



Sensor in free space

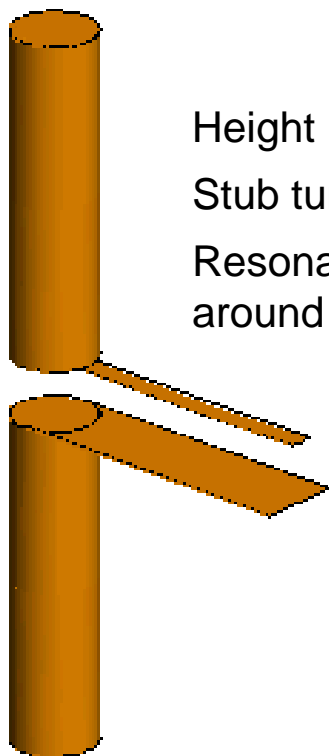


Sensor on cardboard wall

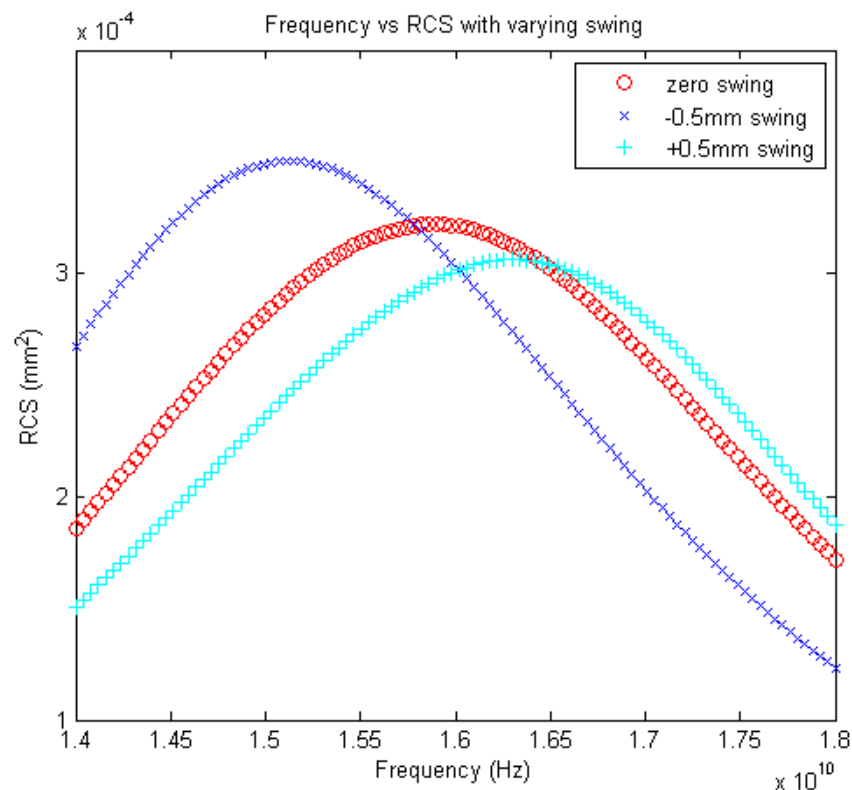
# Dipole Antenna Design



Objective: determine if amplitude or phase modulation dominates.



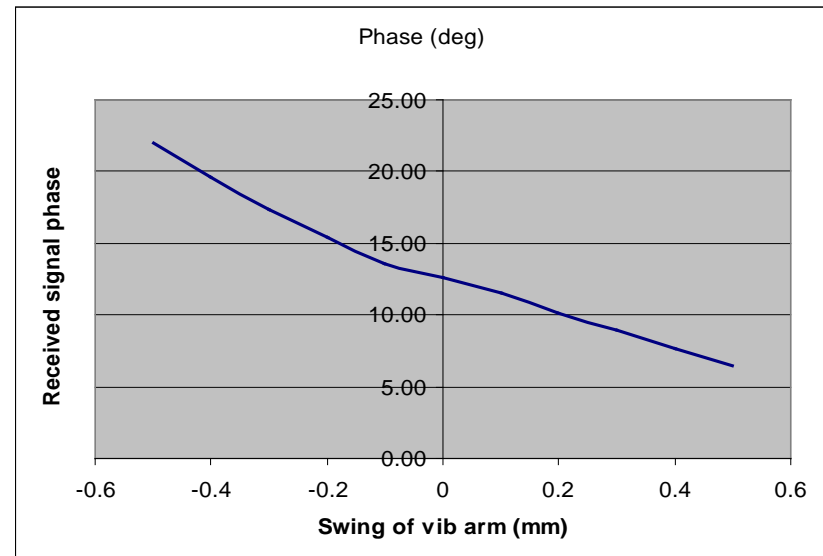
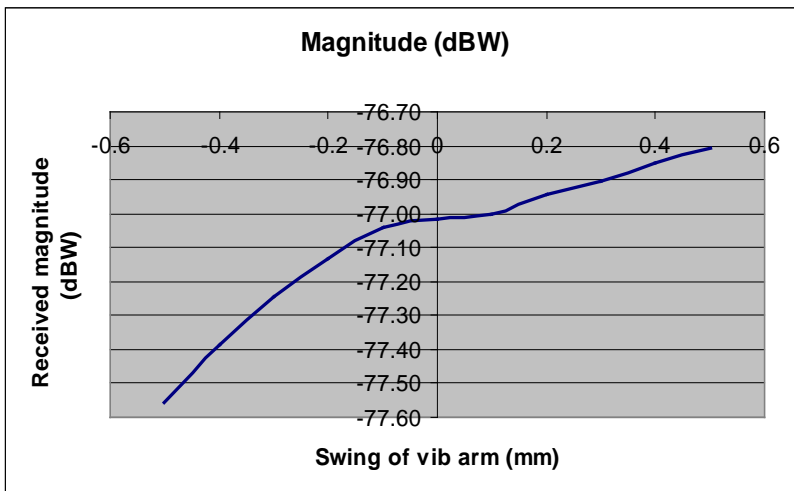
Height 1.12 cm  
Stub tuner: 0.47 cm long  
Resonates electrically  
around 16 GHz



# Dipole antenna

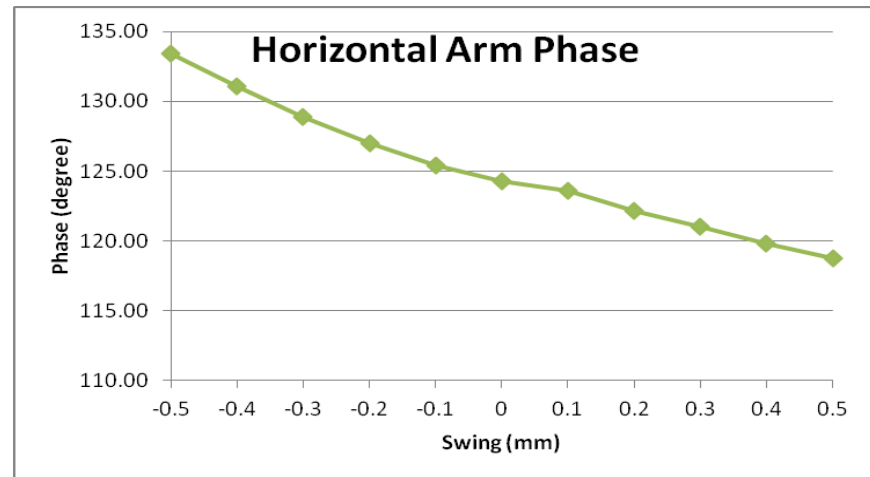
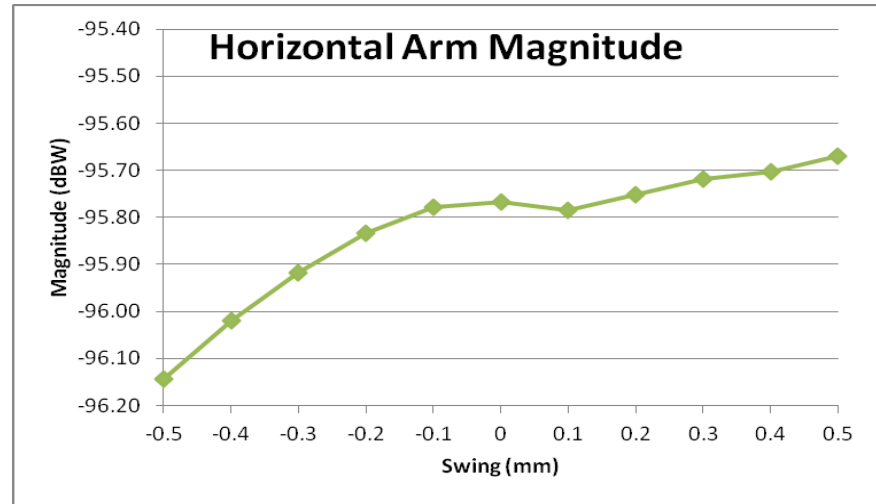
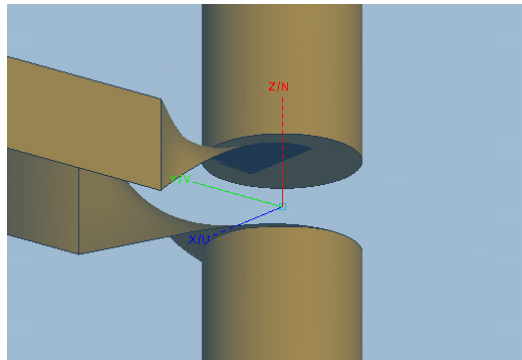
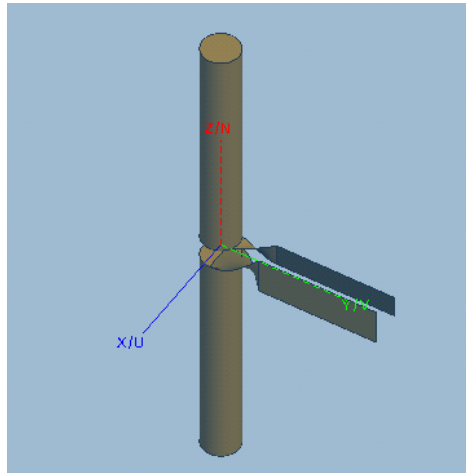


- Assumes interrogator uses horn antenna with 15 dBi gain and 2W output.
- Dipole has low radar cross-section, and therefore low power reflection.



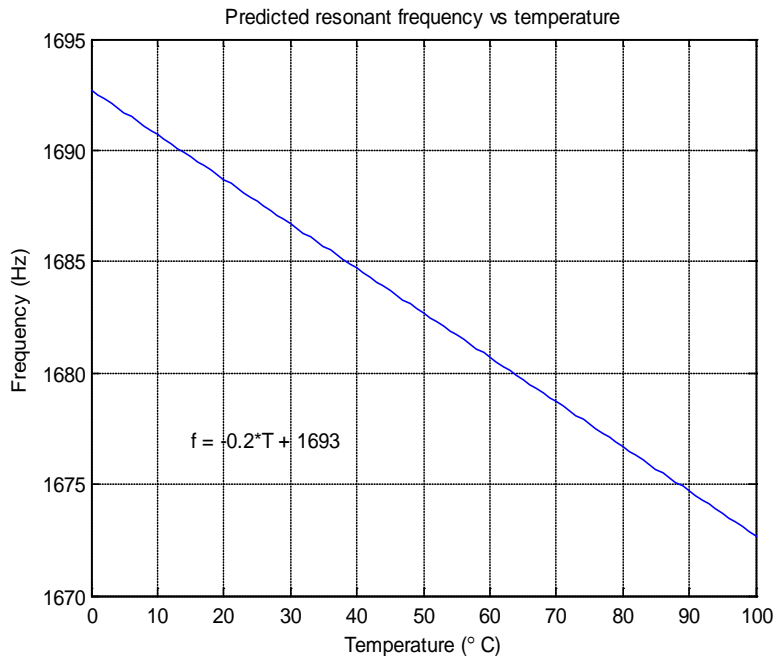


# Dipole w/ rotated tuning arms

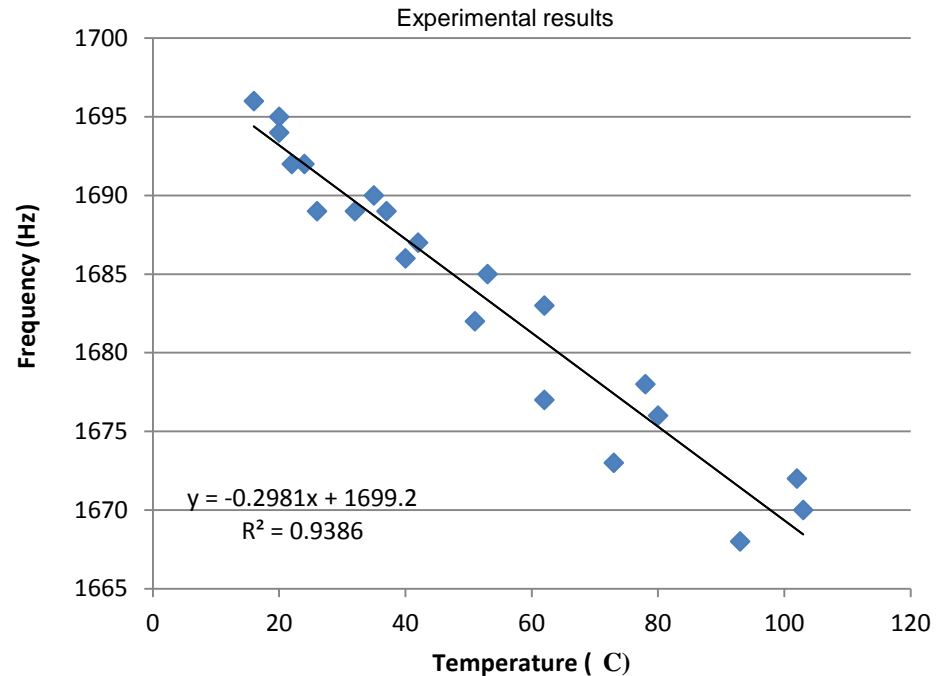


# Temperature dependence

- Elastic modulus, density, and dimensions all depend on temperature and affect natural frequency.
- Modeling vibrating cantilever beam predicts 0.2 Hz/ C



- Experimentally measured peak resonant frequency with changing temperature.
- Data shows 0.3 Hz/ C temperature dependence.

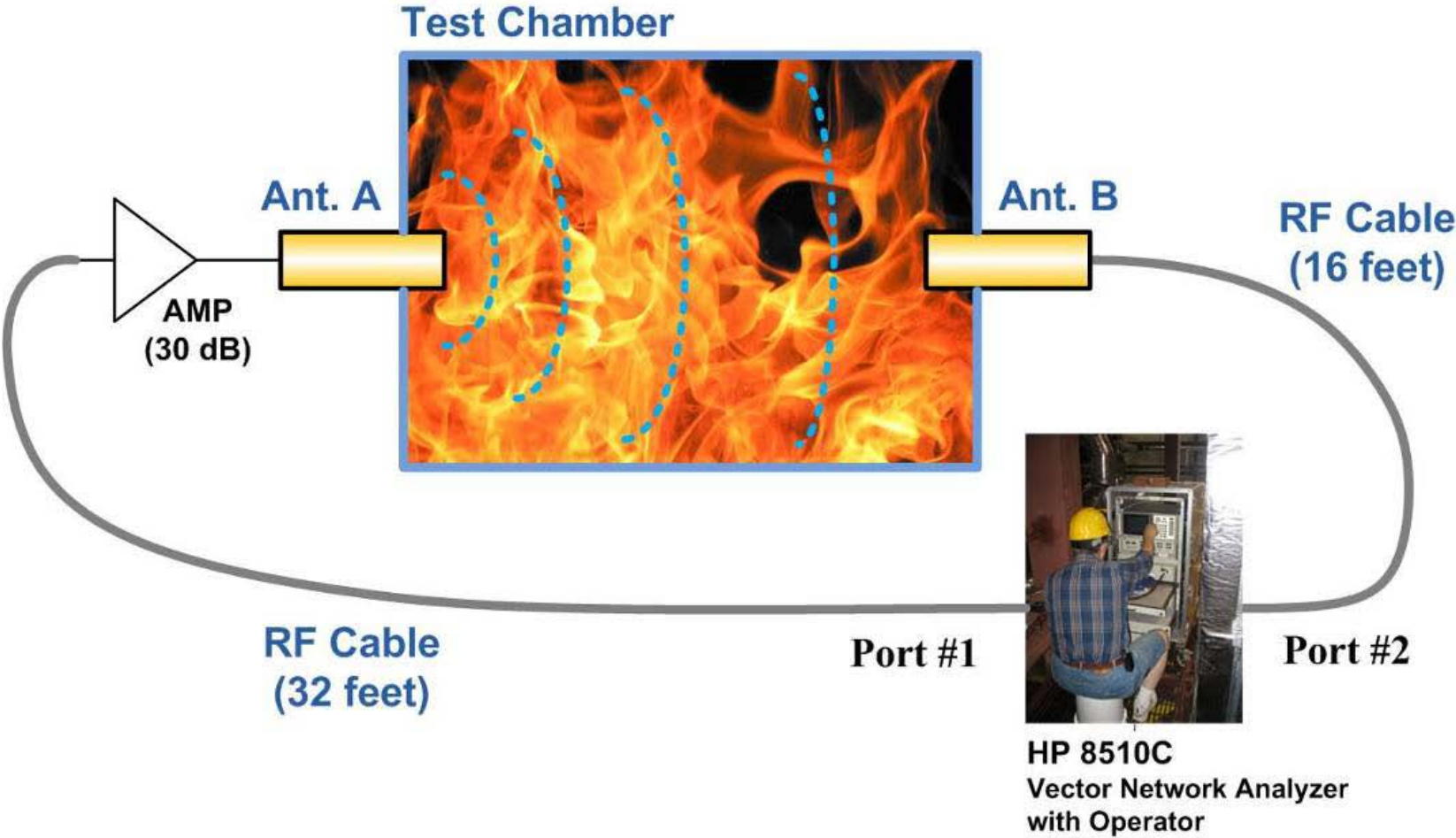


# Characterization of RF Environment

## Overview



Goal: find regions of RF spectrum where attenuation is lowest

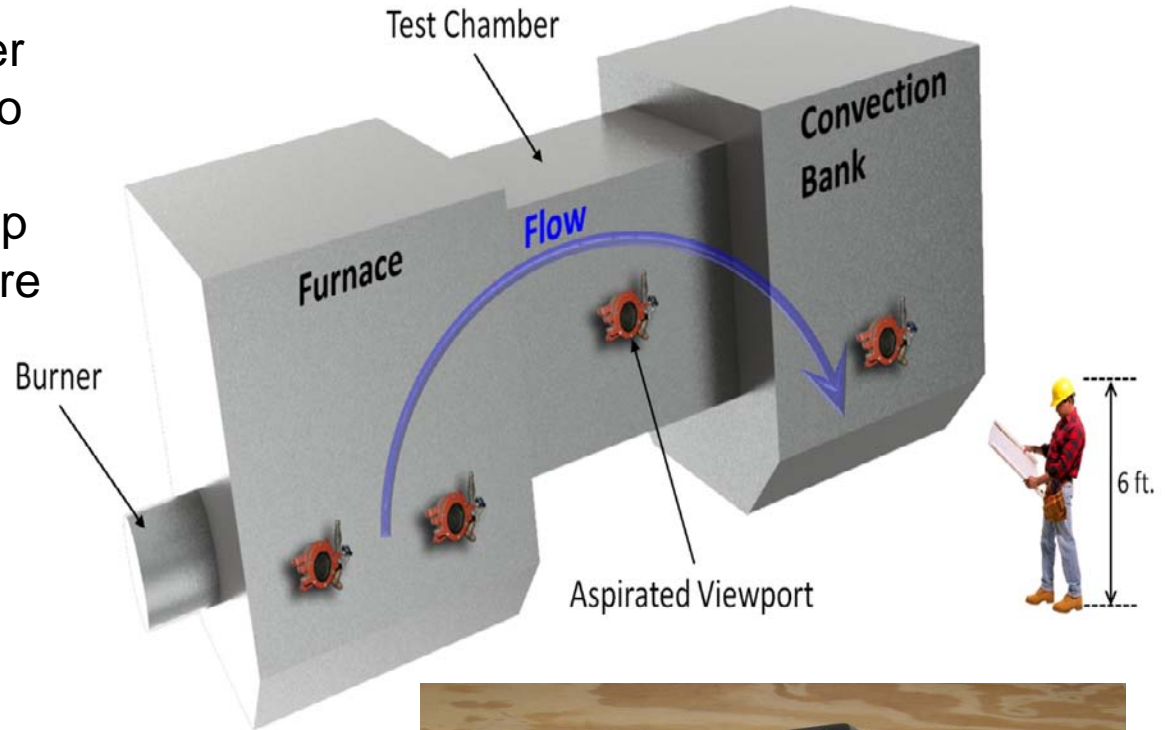


# Characterization of RF Environment

## Site Survey



- Babcock & Wilcox Small Boiler Simulator (SBS), Alliance Ohio
- Tim Fuller and Tom Flynn – B&W Power Generation Group
- B&W IRAD supports cost share

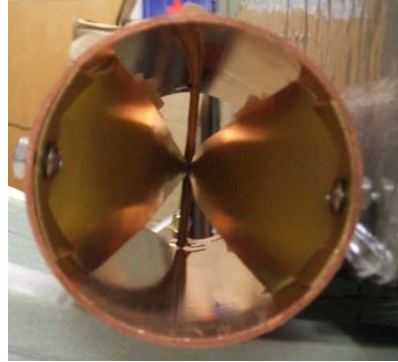


# Characterization of RF Environment

## *Antenna installation*



- antennas need to be broad band (2.5 – 40 GHz)
- must fit inside standard 3" viewport
- needs to withstand high temperatures



Viewpoint installations

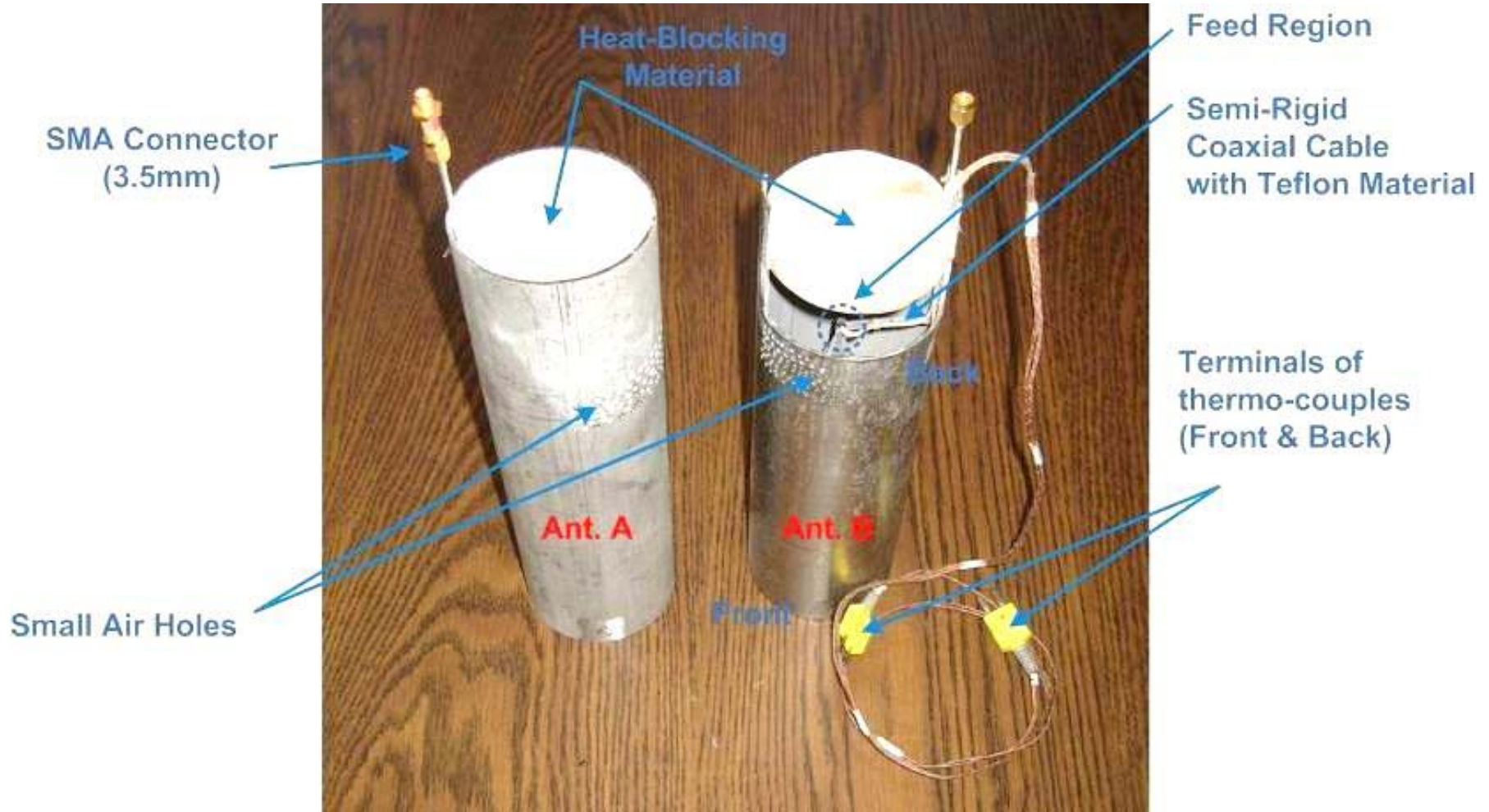


TEM horn antenna



# Characterization of RF Environment

## *Antenna installation*

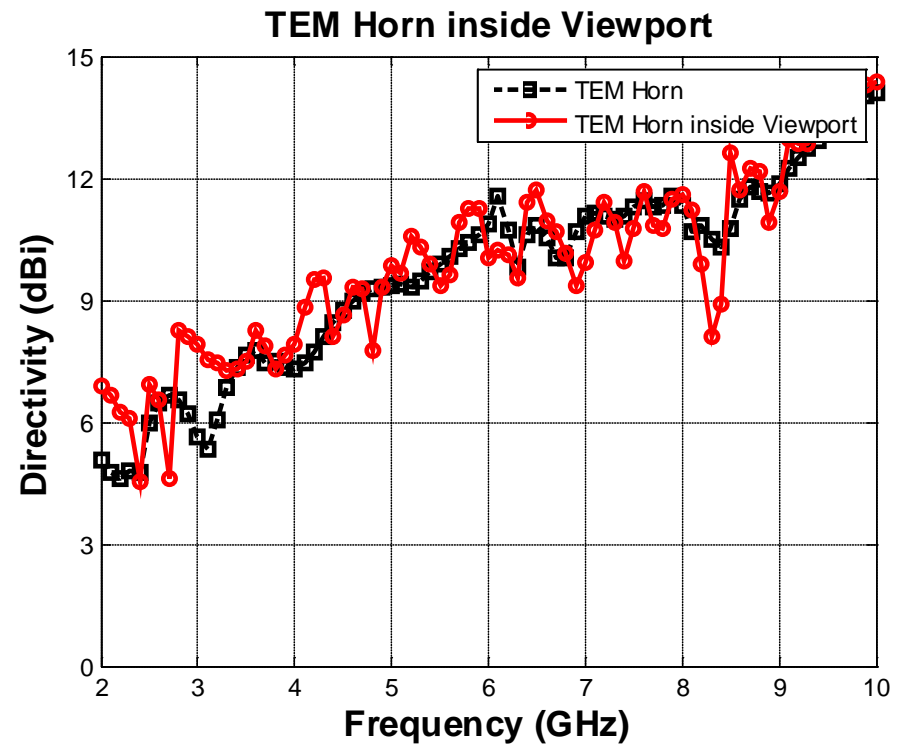
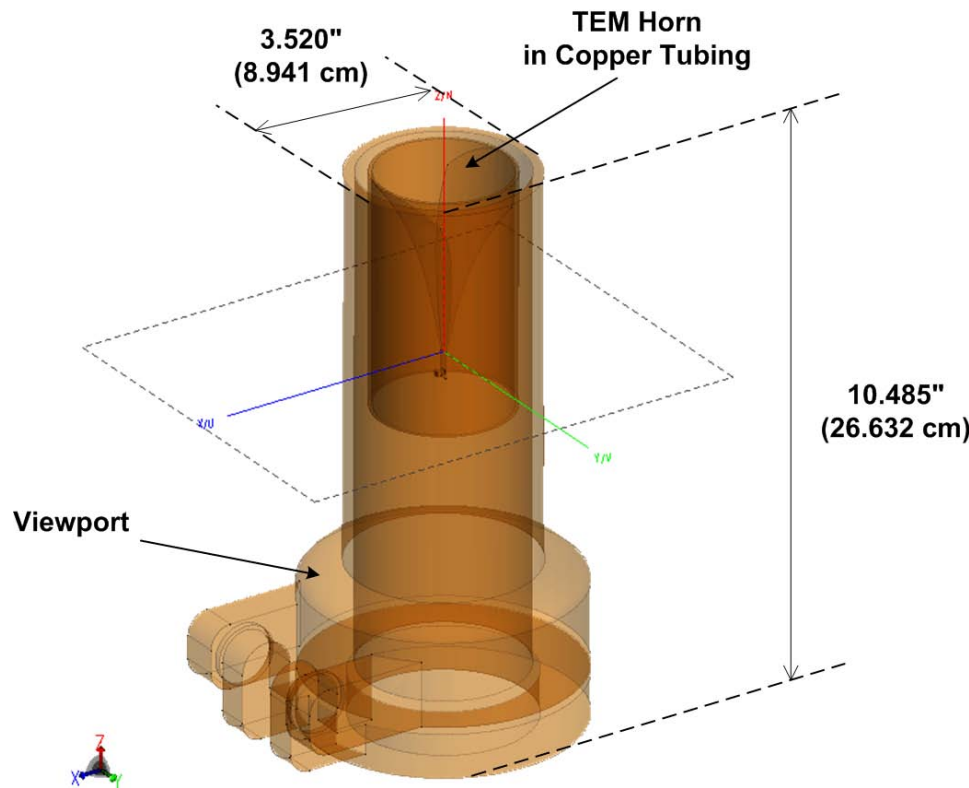


# Characterization of RF Environment

## *Antenna design*



- Viewport installation does not affect gain performance significantly



# Characterization of RF Environment

## *Field Tests*



- 2 field tests
- Natural gas and biomass firing
- SBS convection pass
- Evaluated RF attenuation from 2.5 – 40 GHz



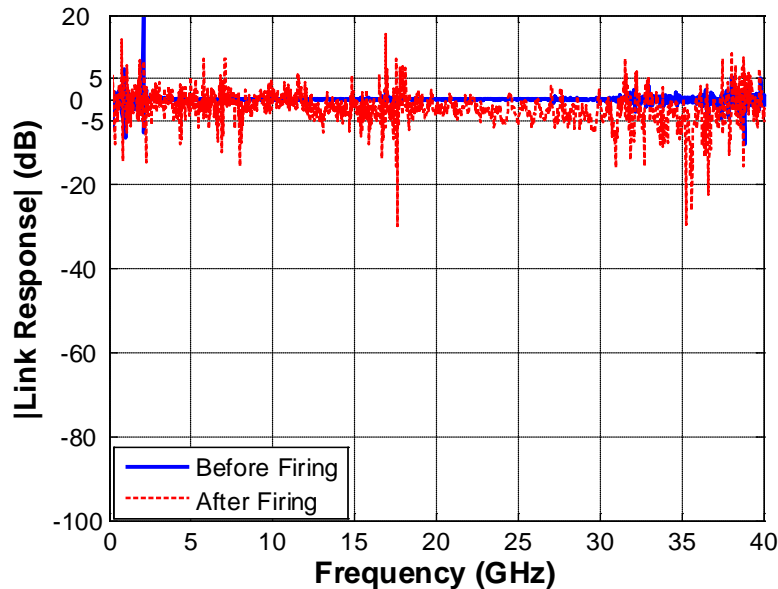


# Characterization of RF Environment

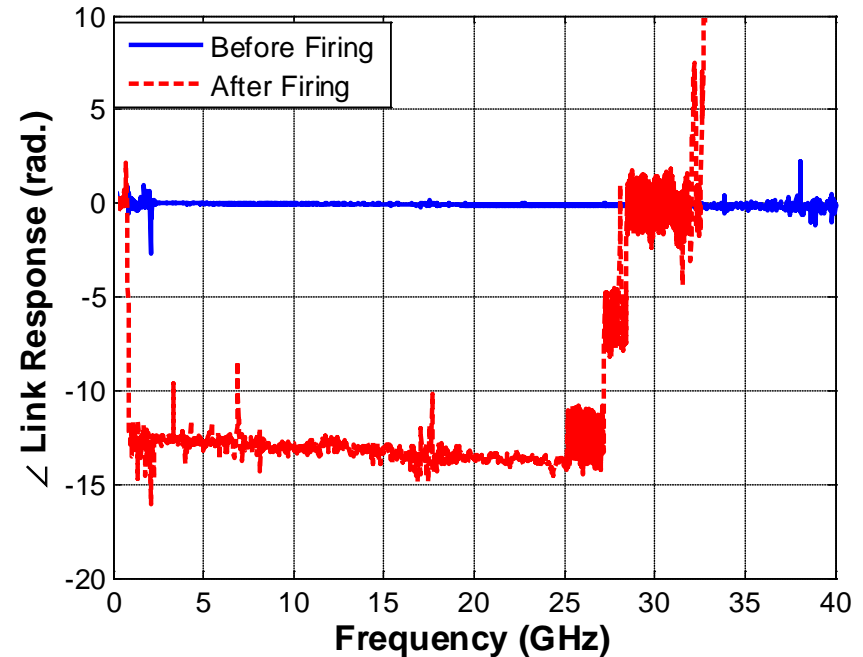
## *Experimental results - biomass*



### Magnitude



### Phase



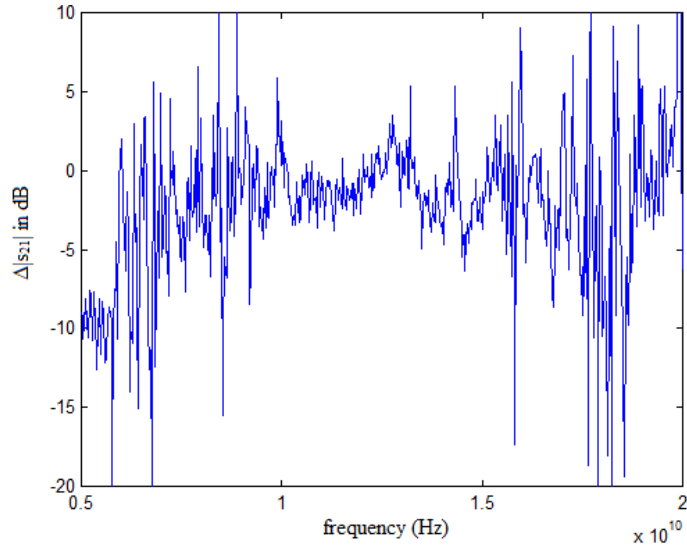
Plots show difference in scattering parameter measurements made before and after biomass firing.

# Characterization of RF Environment

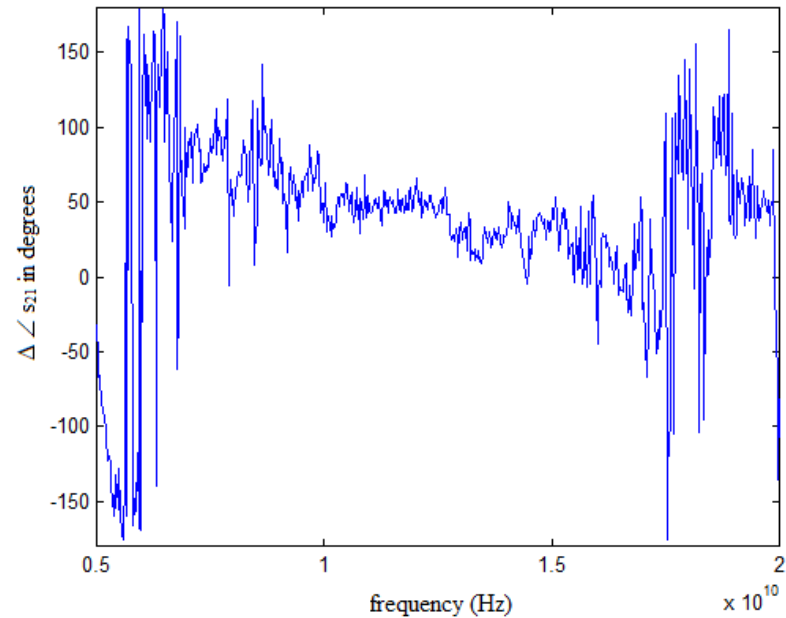


## *Experimental results – coal firing*

Magnitude



Phase



Conclusion: 10 GHz to 15 GHz should be a good window for operation of sensor.

# Conclusions

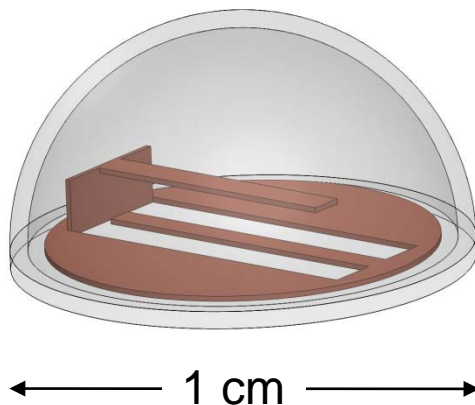


- Testing at B&W SBS indicates RF link budget is satisfactory (natural gas, biomass, and coal firing)
- High temp RF materials characterization –strong response demonstrated (not important to new MMA design)
- Metamaterial sensor concept was not demonstrated – ***indicates strong need for a sensor signal which is distinct from broadcast energy***
- Novel hybrid sensor concept developed – mechanically modulated antenna (MMA) with acoustic excitation
- MMA shows good signal discrimination against background- unique time based signature
- Remote excitation and interrogation experiments successful
- Demonstrated temperature dependent shift in modulation frequency
- MMA sensor shows broad field of view (+/- 70°)
- Modeled modulation mechanism and showed strong phase response

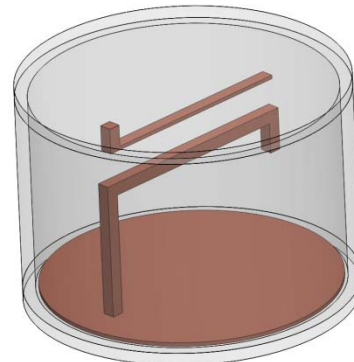
# Next Steps



- Optimize antenna design: maximize change in impedance due to a unit displacement
- Investigate retroreflecting antenna array
- Optimize acoustic capture efficiency; enables lower acoustic power or larger beam deflections
- Evaluate dielectric antenna concept (smaller form factor, sapphire as antenna and package, not influenced by mounting specifics)
- Investigate sensor multiplexing (acoustic and RF multiplex)
- Develop Doppler interrogator (possibly COTS components)
- Develop high-temperature packaging
- Field test of MMA sensors in coal combustion and/or gasifier facilities



Mechanically  
Modulated  
Dielectric  
Antennas



# Contact Information

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